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# Review of Narrowband Impedance-Matching Limitations

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Many antennas can be characterized by their radiation  $Q$  (the ratio of reactance to radiation resistance). A classic problem is determining the maximum possible bandwidth,

$$B_n = (f_{high} - f_{low}) / \sqrt{f_{high} f_{low}},$$

constrained by the maximum permissible reflection magnitude,  $R$ , and the number of tuned circuits,  $n$ , in the impedance-matching circuit. Figure 1 presents a schematic diagram for multiple-tuning networks, for an example antenna characterized by a series combination of capacitance and resistance. The tuned circuits are synchronous. Typically,  $B_n$  is less than one-half (narrowband), and the  $QB_n$  product is less than three.

Bode [1] determined the relationship between  $QB_n$  and  $R$  for the case of an infinite number of tuned circuits:

$$QB_\infty = \frac{\pi}{\ln\left(\frac{1}{R}\right)} \tag{1}$$

Fano [2] determined the relationship between  $QB_n$  and  $R$  for any number of tuned circuits. His solution was in terms of a set of simultaneous equations:

$$QB_n = \frac{2 \sin\left(\frac{\pi}{2n}\right)}{\sinh(a) - \sinh(b)}, \tag{2}$$

$$\frac{\tanh(na)}{\cosh(a)} = \frac{\tanh(nb)}{\cosh(b)}, \tag{3}$$

$$R = \frac{\cosh(nb)}{\cosh(na)}. \tag{4}$$

MATCAD computer software was used to solve this set of equations. Fano's Figure 19 is reproduced in Figure 2.

Lopez [3] had, as an objective, the quantification of the improvement factor when the impedance-matching circuit is increased by one tuning level. He developed the following equations:

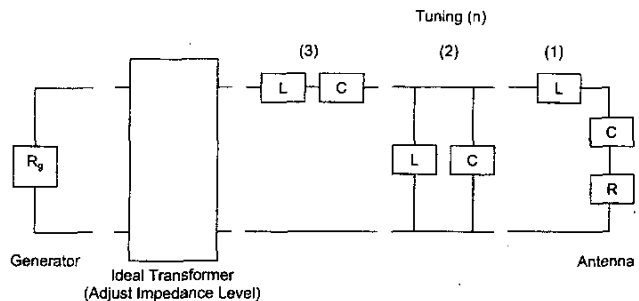


Figure 1. A multiple-tuning impedance-matching network (contiguous tuning elements alternate between series and parallel elements; all tuning elements are synchronous).

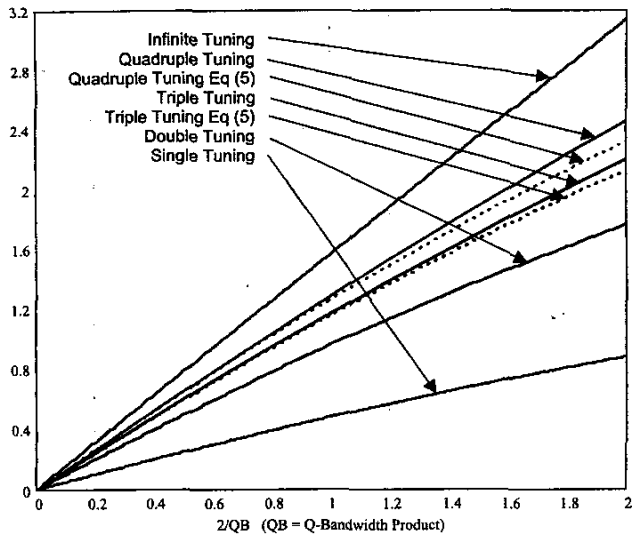


Figure 2.  $\ln(1/R)$  (vertical axis) as a function of  $2/QB$  (Fano's [2] Figure 19).

Table 1. The bandwidth increase for increased tuning,  $R > 1/3$ ,  
 $B_n/B_{n-1} = a_n/a_{n-1}$ .

Number of Tuned Circuits ( $n$ )	$a_n$		Percent Bandwidth Increase $[(B_n/B_{n-1}) - 1]100$
	January 1973	June 2004 (based on Fano)	
1	1	1	
2	2	2	100
3	2.33	2.41	20
4	2.67	2.63	9.1
5	2.76	2.76	4.6
6	2.84	2.84	3.3
7	2.89	2.90	1.8
8	2.93	2.94	1.7
$\infty$	$\pi$	$\pi$	

Table 2. The coefficients  $a_n$  and  $b_n$ .

$n$	$a_n$	$b_n$
1	1	1
2	2	1
3	2.413	0.678
4	2.628	0.474
5	2.755	0.347
6	2.838	0.264
7	2.896	0.209
8	2.937	0.160
$\infty$	$\pi$	0

$$QB_n = \frac{1}{\sinh\left[\frac{1}{a_n} \ln\left(\frac{1}{R}\right)\right]}, \quad (5)$$

$$QB_n \approx \frac{a_n}{\ln\left(\frac{1}{R}\right)} \text{ for } R > \frac{1}{3}. \quad (6)$$

Equation (5) is in exact agreement with the Fano relationship for the cases of  $n=1$  and  $n=2$ . Figure 2 shows the difference between the Fano relationship and Equation (5) for the triple- and quadruple-tuning cases. Equation (6) is in exact agreement with the Fano relationship for all values of  $n$  as  $R$  approaches 1. An initial set of constants,  $a_n$ , was defined in [3], and recently revised when the Fano equations were solved using *MATCAD*. Table 1 presents the old and the new values for  $a_n$ , and the bandwidth-increase factor as  $n$  increases. It is clear that double tuning provides a substantial increase in bandwidth. It is not clear that a 20% increase in bandwidth for triple tuning will outweigh the added complexity of another level of tuning.

Equation (5) has been revised for improved accuracy:

$$QB_n = \frac{1}{b_n \sinh\left[\frac{1}{a_n} \ln\left(\frac{1}{R}\right)\right] + \frac{(1-b_n)}{a_n} \ln\left(\frac{1}{R}\right)}. \quad (7)$$

The coefficients are presented in Table 2.

In 1951, Tanner [4] interpreted Fano's work, and presented design curves that were more suitable for practical antenna engineers. Jasik [5] published these curves in the first edition of his handbook. Unfortunately, these curves do not include the single- and double-tuning cases. Figure 3 adds these cases to the Tanner design chart, and it also provides improved accuracy for triple and quadruple tuning.

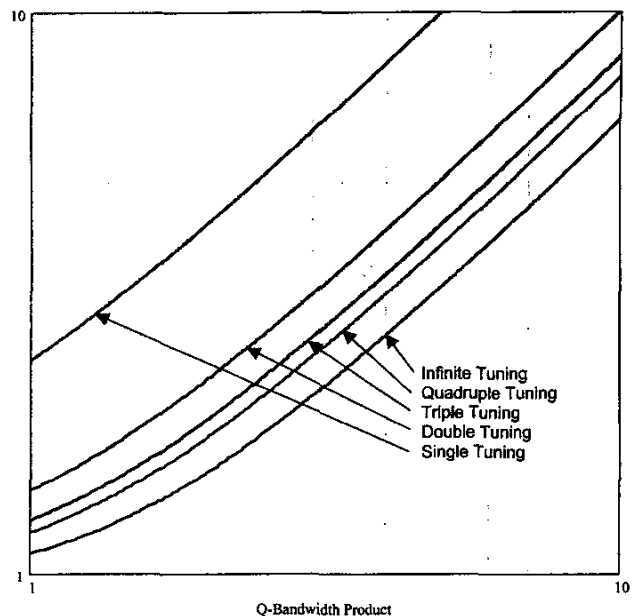


Figure 3. The maximum standing-wave ratio (vertical axis) as a function of  $QB$  (Tanner [4]).

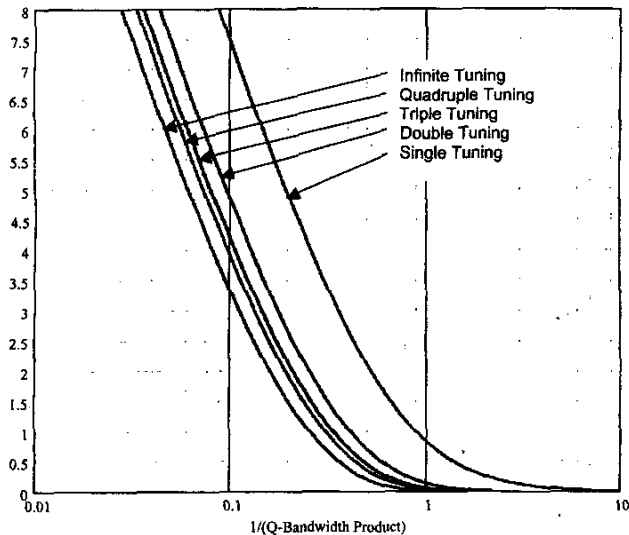


Figure 4. The maximum impedance mismatch loss (in dB, vertical axis) as a function of  $1/QB$  (Matthaei, Young, and Jones [6], Figure 4.09-3).

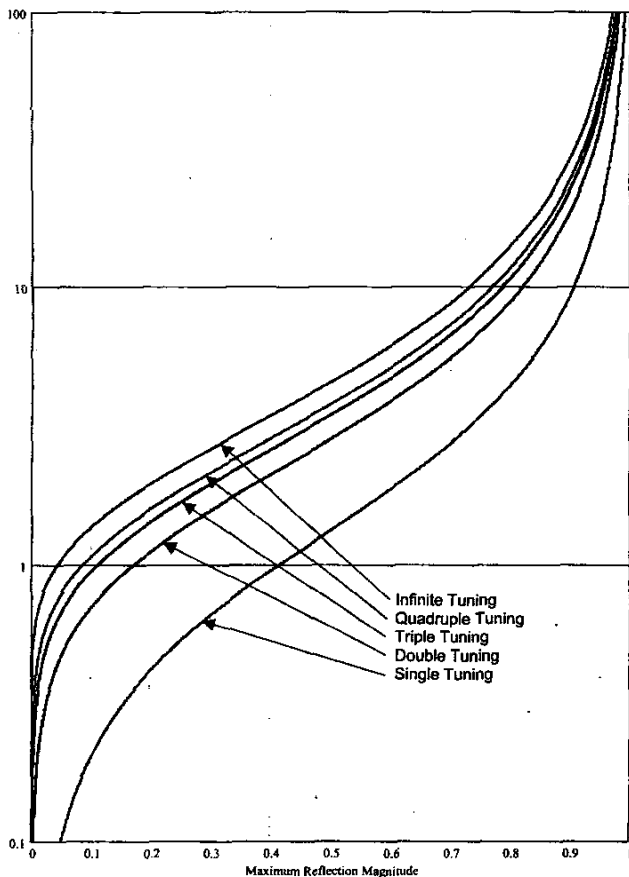


Figure 5.  $QB$  (vertical axis, logarithmic scale) as a function of  $R$  (Wheeler [7], Figure 3).

Matthaei, Young, and Jones [6] published curves of maximum impedance-mismatch loss in dB versus  $1/Q$ . They normalized the bandwidth to one. The horizontal scale can be relabeled as  $1/QB$ . Equation (7) was used to reproduce their results, which are presented in Figure 4.

Wheeler [7] published curves of  $QB$  versus  $R$  for single, double, and infinite tuning. Figure 5 includes triple and quadruple tuning for this presentation. The scale for the  $Q$ -bandwidth product was changed to increase the range.

The derivation of the relationships in [3] was based on the teaching of Harold A. Wheeler [8, 9, 10].

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### Ideas for Antenna Designer's Notebook

Ideas are needed for future issues of the Antenna Designer's Notebook. Please send your suggestions to Tom Milligan and they will be considered for publication as quickly as possible. Topics can include antenna design tips, equations, nomographs, or shortcuts, as well as ideas to improve or facilitate measurements. ☺